The Magnetic Field of Current-Carrying Polygons: An Application of Vector Field Rotations

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The magnetic field of current-carrying polygons: An application of vector field rotations

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We calculate the magnetic field around a current loop consisting of a regular N-sided polygon by successively rotating the field obtained from a straight, finite, current-carrying wire. This involves developing and applying a vector field rotation operator, which transforms vector fields the way a rotation matrix transforms scalar fields. Using this result, we explore the various magnetic field components about the current loop and notice an interesting structure that results from the geometry of the polygons. The magnetic field on the axis of the current loop as well as the field at all points around a circular loop are considered as limiting cases. © 2000 American Association of Physics Teachers.

I. INTRODUCTION

A standard calculation that often appears in an introductory physics course is the magnetic field due to a constant current flowing through a finite length of wire. Students are then asked in homework problems to find the magnetic field of a current arrangement at some point of symmetry, such as the center of a square or a circle. Even in more advanced treatments of electromagnetism, graduate school that a student finally witnesses the full calculation of the magnetic field due to a circular current loop. The approach used here is conceptually not much different than the calculation for a finite wire, thus providing a relatively straightforward, albeit somewhat tedious, method for obtaining the field of a current-carrying polygon or circle.

We begin with the magnetic field due to a finite length of wire. This is written most simply as

$$B = \frac{\mu_0 i}{4\pi d} (\sin x_2 - \sin x_1),$$

with a direction given by the right-hand rule. Here, $i$ is the current, $d$ is the length of the perpendicular segment from the axis of the wire to the observation point $P$, and $x_1$ and $x_2$ are the angles formed between this perpendicular segment and lines drawn from the observation point to the ends of the wire. This is shown in Fig. 1. If the length of the current segment is taken to be $2a$ and we orient it parallel to the $y$ axis with its center a distance $R$ along the $x$ axis (see Fig. 1), Eq. (1) becomes

$$B = \frac{\mu_0 i}{4\pi d} \left( \frac{y + a}{\sqrt{d^2 + (y + a)^2}} - \frac{y - a}{\sqrt{d^2 + (y - a)^2}} \right) \times \left( \frac{z}{d} \hat{e}_x + \frac{x - R}{d} \hat{e}_z \right),$$

where $d = \sqrt{(x - R)^2 + z^2}$, and the direction of the field is now explicitly written.

II. MAGNETIC FIELD DUE TO CURRENT LOOPS OF REGULAR POLYGONS

To obtain the magnetic field of a current loop in the shape of an $N$-sided polygon, we simply need to add up the contributions from $N$ different current segments arranged in the shape of a regular polygon. To sum the fields in a straightforward manner, imagine introducing a new coordinate system, $x'y'$, that is rotated by an angle $\theta = 2\pi/N$ with respect to the $xy$ system. This will place the second segment of the polygon in an identical position in the primed coordinate system as the first segment is in the unprimed coordinate system. Thus, the magnetic field of this new segment will be given by Eq. (2), with $x \rightarrow x'$, $\hat{e}_x \rightarrow \hat{e}_x'$, $y \rightarrow y'$, etc. That is, the field of the second segment will look identical in the primed coordinate system to the field of the first segment in the unprimed system. Similarly, one can imagine a set of $N$ coordinate systems rotated by angles $\theta_n = 2\pi n/N$ (for $0 \leq n \leq N - 1$), each of which gives rise to an identical formula for the magnetic field in reference to a different coordinate system. The total magnetic field would then be just the sum of these,

$$B_{tot} = B + B' + B'' + \cdots + B''''',$n

where the final term (with $N - 1$ primes) refers to the magnetic field of the $N$th segment in the $N$th coordinate system. Of course, since this result references $N$ different coordinate systems, it is not all that useful in this form. Therefore, we must rotate each term in Eq. (3) so that they all refer to the same coordinate system. This is more subtle than it sounds because it involves rotating a vector field. Although this sounds simple, the traditional “coordinate system rotation” does not completely accomplish this task. Because this topic is not typically covered in undergraduate texts, we will give a fairly detailed description in the following subsections.

A. Rotating a scalar field

To begin, let us consider a coordinate system, $x'y'z'$, that has been rotated about the $z$ axis by an angle $\theta$ with respect to a second coordinate system, $xyz$. The relationship between points in the primed system, $x'$, and those in the unprimed system, $x$, is given by $x' = Rx$, where
Fig. 1. A finite wire segment with constant current \( i \) is shown with an appropriate coordinate system. The magnetic field of this segment is given by Eq. (1), where use of the right-hand rule is implied, and also in Cartesian coordinates by Eq. (2).

The primed coordinates as

Viewed in the unprimed coordinates, each of the transformed
implies, as we claimed earlier, that
function under this ‘coordinate system rotation.’ For ex-
thing we would like to consider is what happens to a scalar
z
A
it is tempting to say that
A
A
\(~ x \\hat{e}_z \) looks in the unprimed system, this is not the
function, \( A(x) = x \) looks the same in
planes, \( x = b \), it contains vectors with the same magnitude and direc-
the scalar function, \( A_e(x) = x \), rotates to give \( A(x') = A(Rx) = (x \cos \theta + y \sin \theta \hat{e}_x \), and hence the new equifield
logically, the equifield lines have been rotated exactly the same
way as in the previous example. Note, however, that these
vectors are still pointing in the \( \pm x \) direction. To accomplish
true rotation of the vector field requires that we spin the
vectors by an angle \( \theta \) in addition to rotating their positions.

To spin the vectors, we appeal to the rotation matrix once
again and consider \( A'(x) = RA(x) \). This will spin every vec-
tor in a direction opposite to that of a function under a coordi-
ate rotation \( x' = Rx \). Since we want to spin the vectors in the
same direction as their positions rotate, we must trans-
form according to \( A'(x) = \mathcal{R}(-\theta)A(x) = \mathcal{R}^{-1}(\theta)A(x) \). For
an arbitrary vector field, the transformed components are then
given by \( (A'_x , A'_y , A'_z ) = (A_x \cos \theta - A_y \sin \theta , A_x \sin \theta + A_y \cos \theta , A_z ) \). Returning once again to our example, \( A(x) = x \hat{e}_x \), we have \( (A'_x , A'_y , A'_z ) = (x \cos \theta \sin \theta , x \sin \theta , 0) \). This new
vector field still has equifield lines given by \( x = b \), indicating that
the locations of the vectors have not changed. The com-
ponents of these vectors, however, are given by \( b \cos \theta , b \sin \theta \), demonstrating that they have been spun by
an angle \( \theta \), as desired.

C. Rotating a vector field

In the last two subsections, we have seen that the rotation
matrix \( \mathcal{R} \) can be applied to a vector field in two different
ways. Rotating the positions of the vectors by an angle \( \theta \) is
accomplished via \( A' = \mathcal{R}^{-1}A \). Rotating a vector field is then achieved by com-
bing these two operations, which defines a vector field ro-
tation operator, \( \mathcal{R}(\theta) \), by

\[
A'(x') = \mathcal{R}(\mathcal{A}(x)) = \mathcal{R}^{-1}(\theta)A(Rx).
\]  

(6)

Equation (6) is valid for rotations about any axis provided
the rotation matrix is changed accordingly. For rotations
about the \( z \) axis, the component functions are given by

\[
A'_x = \cos \theta A_x (x \cos \theta + y \sin \theta, -x \sin \theta + y \cos \theta, z)
- \sin \theta A_y (x \cos \theta + y \sin \theta, -x \sin \theta + y \cos \theta, z),
A'_y = \sin \theta A_x (x \cos \theta + y \sin \theta, -x \sin \theta + y \cos \theta, z)
+ \cos \theta A_y (x \cos \theta + y \sin \theta, -x \sin \theta + y \cos \theta, z),
A'_z = A_z (x \cos \theta + y \sin \theta, -x \sin \theta + y \cos \theta, z).
\]  

(7)

Using our example function one last time, the rotated com-
ponents of \( A(x) = x \hat{e}_x \) are given by \( A'_x = (x \cos \theta + y \sin \theta \cos \theta, A'_y = (x \cos \theta + y \sin \theta \sin \theta, and A'_z = 0 \). As
before, the equifield lines are given by \( y = -x \cot \theta + b \csc \theta \), indicating that the positions have been rotated.
In addition, the components of the vectors on these lines are

rotated position. Thus, a coordinate rotation does not result in a true rotation of a vector field.

Let us extend our previous example and consider the vec-
tor field \( A(x) = x \hat{e}_x \). This function has all vectors pointing in
the \( \pm x \) direction. In addition, each of the ‘equifield’ lines,
\( x = b \), contains vectors with the same magnitude and direc-
tion.

B. Moving and spinning a vector field

Next, let us consider what happens to a vector function,
\( A(x) \), subjected to a coordinate rotation, \( x' = Rx \). Although
it is tempting to say that \( A(x') \) looks identical in the primed
system as \( A(x) \) looks in the unprimed system, this is not the
case. Since each component of \( A(x) \) is a scalar function, we have

\[
A(Rx) = A_x(Rx) \hat{e}_x + A_y(Rx) \hat{e}_y + A_z(Rx) \hat{e}_z .
\]  

(5)

Viewed in the unprimed coordinates, each of the transformed
scalar functions in Eq. (5) has been rotated by an angle \( \theta \)
with respect to the original functions. Notice, however, that
the unit vectors have not changed. This means the compo-
nents of each transformed vector are exactly the same as
the components of the original vector. To put it another way,
the vector itself has not changed, it has simply moved to a new
Armed with this vector field rotation operator, each term in Eq. (3) can now be written in terms of a single coordinate system. The result is

\[
\mathbf{B}_{\text{tot}}(x) = \sum_{n=0}^{N-1} \mathcal{R}_n^{-1} \mathbf{B}(\mathcal{R}_n x),
\]

(8)

where the subscript \(n\) designates a rotation angle of \(\theta_n = \frac{2\pi n}{N}\). This formula sums the magnetic field contributions from \(N\) current-carrying segments of a regular polygon centered at the origin. The field of a single segment of a polygon is obtained from Eq. (2) by including a geometric constraint between the length of a segment and the “radius” of the polygon given by \(a = R \tan(\pi/N)\). Inserting this single-segment field into Eq. (8) yields the final result,

\[
\mathbf{B}_{\text{tot}} = \sum_{n=0}^{N-1} \frac{\mu_0 i}{4\pi R} \left[ \tan\left(\frac{\pi}{N}\right) + \frac{(x \sin \theta_n - y \cos \theta_n)}{H_+} + \frac{(x \sin \theta_n - y \cos \theta_n)}{H_-} \right]
\]

\[
	imes \left[ \frac{z \cos \theta_n \hat{e}_x + z \sin \theta_n \hat{e}_y + (1 - x \cos \theta_n - y \sin \theta_n) \hat{e}_z}{(1 - x \cos \theta_n - y \sin \theta_n)^2 + z^2} \right],
\]

where

\[
H_+ = \sqrt{x^2 + y^2 + z^2} \tan\left(\frac{\pi}{N}\right) - 2(x \cos \theta_n + y \sin \theta_n) \pm 2(x \sin \theta_n - y \cos \theta_n) \tan\left(\frac{\pi}{N}\right)
\]

(9)

are the distances from the observation point to the ends of the wire segments and \(r = \sqrt{x^2 + y^2 + z^2}\) is the distance of the observation point from the origin. Here, and for the remainder of this paper, we use dimensionless coordinates \(x, y,\) and \(z\) that have been scaled by the radius of the polygon, \(R\).11

Equations (9) and (10) represent the exact result for the magnetic field of a current loop in the shape of a regular, \(N\)-sided polygon centered at the origin. Unfortunately, since each term in the sum is equally important, this result is not very convenient for analytic calculations. However, modern numerical analysis programs such as Mathematica or Maple can handle equations such as these with relative ease. As an example, Fig. 2 shows a series of constant \(z\) contour plots of \(B_x, B_z,\) and \(|B|\) for a hexagon. In this figure, lighter regions represent larger (positive) values and darker regions correspond to smaller (or more negative) values. Because subtle
features of the field are difficult to discern by relying solely on the coloring scheme, we have included some contour lines on these graphs. Since the range of field values decreases dramatically for increasing \( N \), we have used two different sets of equally spaced contour lines, one for the \( z = 0 \) and \( z = 0.1 \) plots, and another, with a spacing about seven times smaller, for the \( z = 0.5 \) and \( z = 1.0 \) plots. These two sets of contour lines are superimposed on the color scale.

A surprising feature in this figure is the appearance of “islands” of local maxima and minima in the \( B_z \) and \( |B| \) plots. This behavior is present in all polygons, but becomes less and less pronounced as \( N \) increases. Explaining exactly why these islands appear will challenge even the brightest students. Because of these extrema, it can be a bit confusing to try and match up the contour lines on the graphs with those on the color scale. For example, there are six curves on the graph of \( |B| \) for \( z = 0 \) even though the color scale shows only three positive contour lines. This is because there is a maximum (on top of the wire) on this graph that results in two contour lines for each of the magnetic field values chosen. Also, it is worth mentioning that the outermost curves on the graphs of \( |B| \) for \( z = 0.5 \) and \( z = 1.0 \) correspond to the smallest positive contour line on the color scale.

III. SPECIAL CASES

A. On the axis

The magnetic field on the axis of the current loop is easily found by setting \( x = y = 0 \) in Eq. (9) to get

\[
B_p = \sum_{n=0}^{N-1} \frac{\mu_0 i}{4 \pi R} \frac{2 \tan (\pi N)}{\sqrt{\sec^2 (\pi N) + z^2}} \left( \frac{z \cos \theta_n \hat{e}_x + z \sin \theta_n \hat{e}_y + \hat{e}_z}{1 + z^2} \right).
\]

The \( x \) and \( y \) terms sum to zero by symmetry and each of the \( z \) terms are identical, giving

\[
B_p = \frac{\mu_0 i}{4 \pi R} \frac{2N \tan (\pi N)}{\sqrt{\sec^2 (\pi N) + z^2}} \left( \frac{1}{1 + z^2} \right) \hat{e}_z.
\]

It is instructive to compare this result to the magnetic field on the axis of a circular current loop, obtained by taking the limit \( N \to \infty \) in Eq. (12) to get \( B_p = \frac{\mu_0 i}{2 \pi R} \frac{2N \tan (\pi N)}{\sqrt{\sec^2 (\pi N) + z^2}} \). Figure 3 shows the ratio of these fields, \( \frac{B_p}{B_c} \), for \( 3 \leq N \leq 6 \). Notice that the on-axis fields of the polygons are smaller than that of the circle for small values of \( z \). This reflects the fact that points on the polygons are, on average, further from the origin than points on the circle. As \( z \) is increased, the on-axis fields of the polygons become larger than that of the circle and their ratio appears to level off to a constant value. This reflects the dipole behavior of the fields for large \( z \) and the fact that the polygons have a larger area than the circle.

Quantitatively, the fields at the center of the current loops are obtained by taking \( z = 0 \) to get \( B_p/B_c = (N/\pi) \tan (\pi N) \). The large \( z \) behavior is obtained by expanding \( B_p \) and \( B_c \) for \( z \approx 1 \). In this case, the fields can both be written

\[
B = \frac{\mu_0}{4 \pi} \frac{2iA}{(zR)^3},
\]

where \( A \) is the area of the current loop. Note that this is just the on-axis field of a magnetic dipole with dipole moment \( m = iA \). For a circle and polygon, the areas are given by \( A_c = \pi R^2 \) and \( A_p = \pi R^2 \tan (\pi N) \). Thus, we see that as \( z \) increases, the ratio of the fields in Fig. 3 asymptotically approaches the ratio of their areas, \( \frac{B_p}{B_c} \to A_p/A_c = (N/\pi) \tan (\pi N) \).

B. A circular current loop

A more challenging calculation is to find the magnetic field at all points due to a circular current loop. This is accomplished by taking the limit \( N \to \infty \) in Eq. (9). We begin by expanding to lowest order in \( 1/N \), which yields

\[
B \approx \sum_{n=0}^{N-1} \frac{\mu_0 i}{4 \pi R} \left( \frac{z \cos \theta_n \hat{e}_x + z \sin \theta_n \hat{e}_y + (1 - x \cos \theta_n - y \sin \theta_n) \hat{e}_z}{1 + r^2} \right) \times \Delta \theta,
\]

where \( \Delta \theta = 2 \pi/2N \) is the angle subtended by one segment of the polygon. In the limit \( N \to \infty \), this sum becomes an integral,

\[
B = \frac{\mu_0 i}{4 \pi R} \left( \frac{2 \pi z \cos \theta \hat{e}_x + z \sin \theta \hat{e}_y + (1 - x \cos \theta - y \sin \theta) \hat{e}_z}{1 + r^2} \right) \frac{d \theta}{1 + r^2},
\]

giving the magnetic field of a circular current loop in terms of Cartesian coordinates.

Because of the underlying symmetry, it is appropriate to convert this result into cylindrical coordinates \((\rho, \phi, z)\). Substituting for the coordinates and unit vectors, defining \( \omega = \phi - \theta \), and exploiting the periodicity of the trig functions, we end up with

\[
B(\rho, z) = \frac{\mu_0 i}{4 \pi R} \int_0^{2 \pi} \frac{2 \pi z \cos \omega \hat{e}_x + (1 - \rho \cos \omega) \hat{e}_z}{1 + \rho^2 + z^2} d \omega.
\]

As expected, the \( \phi \)-component has integrated to zero. Although a bit tedious, this result can be written in terms of complete elliptic integrals,
where \( k = \sqrt{4\rho((1+\rho)^2+z^2)} \). Equations (17) and (18) represent the magnetic field due to a circular current loop of radius \( R \) centered at the origin. Figure 4 shows some constant \( z \) contour plots of \( B_x \), \( B_z \), and \( |B| \) for a circular current loop. The same shading scheme and contour lines have been used as in Fig. 2 to allow for a direct comparison. One obvious difference is that the islands of extrema that exist for polygons have coalesced into annuli for the circle. This is easily observed by referring to the graphs of \( |B| \) for \( z = 0.1 \) or \( z = 0.5 \). In each case, the small islands in the hexagon plots have given way to an annulus with two contours in each of the circle plots. This is why there are more curves in some of the graphs than there are contours on the color scale. While this same behavior occurs in the \( B_z \) graphs for \( z = 0.1 \) or \( z = 0.5 \), the specific contour lines chosen do not necessarily reveal this behavior.

Clearly, the fields of a polygon and a circle are quite different when you are close to the loop but appear more and more alike the farther you move away. While this is not terribly surprising in light of our discussion regarding Fig. 3, we did not expect to see such a striking similarity at a distance of only one radius.

**IV. CONCLUSION**

We have calculated the magnetic field of a current loop in the shape of a regular, \( N \)-sided polygon centered at the origin. We obtained this result by applying a vector field rotation operator to the magnetic field of a straight, finite, current-carrying wire. This problem has a rich mixture of analytic and computational aspects that make it useful as an advanced undergraduate student project. Alternatively, it would make a nice addition to a course on electromagnetism or mathematical physics, either as part of the lectures or as the basis for a series of homework problems.

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Fig. 4. Contour plots of \( B_x \), \( B_z \), and \( |B| \) for the circular current loop shown in the upper left corner. For comparison, the same color scheme and contour lines have been used as in Fig. 2.
replaced by $\psi(Rx)$, which means that the value $\psi(Rx)$ has moved from $Rx$ to $x$.


We have made the variable substitutions $x' = x/R$, $y' = y/R$, $z' = z/R$ and then dropped the primes.

In general, the contours used for the $B_x$, $B_z$, and $|B|$ plots are the same. However, for $z = 0.5$, the contours used on the $|B|$ plot have a slightly different spacing than those on the $B_x$ and $B_z$ plots. The difference is less than 1% and was chosen to make the local maxima visible.


ANTHROPIIC REASONING

This sort of reasoning is called anthropic, and it has a bad name among physicists. Although I have used such arguments myself in some of my own work on the problem of the vacuum energy, I am not that fond of anthropic reasoning. I would personally be much happier if the values of all the constants of nature could be precisely calculated on the basis of fundamental principles, rather than having to think about what values are favorable to life. But nature cares little about what physicists prefer.